

Model Predictive Control and Reinforcement Learning

– Lecture 1: Dynamic Systems and Simulation –

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universität freiburg



- 1 Dynamic System Models
- 2 From Continuous to Discrete Time
- 3 Input Output Models
- 4 Stochastic Models

Acknowledgement



Slides contain some figures from slides by Rien Quirynen and from the textbook
Model Predictive Control: Theory, Computation, and Design (by Rawlings, Mayne, and Diehl)



- ▶ optimal control = optimization of dynamic systems
- ▶ each optimal control problem (OCP) is characterized by three ingredients:
 - ▶ dynamic system model
 - ▶ constraints
 - ▶ objective function, i.e., cost or reward



- ▶ optimal control = optimization of dynamic systems
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 - ▶ **dynamic system model** (focus of this talk)
 - ▶ constraints
 - ▶ objective function, i.e., cost or reward



- ▶ system model describes evolution of system as function of
 - ▶ system **state** s from state space $\mathbb{S} \subset \mathbb{R}^{n_s}$ (or $\subset \mathbb{Z}^{n_s}$ for discrete states)
 - ▶ control **action** a from action space $\mathbb{A} \subset \mathbb{R}^{n_a}$ (or $\subset \mathbb{Z}^{n_a}$ for discrete actions)
 - ▶ random **disturbance** ϵ from some disturbance space \mathbb{D}

- ▶ examples:

- ▶ **stochastic discrete time system**, for $k = 0, 1, 2, \dots$

$$\boxed{s_{k+1} = f(s_k, a_k, \epsilon_k)} \quad \text{with "evolution function" } f : \mathbb{S} \times \mathbb{A} \times \mathbb{D} \rightarrow \mathbb{S}$$

- ▶ deterministic continuous time **ordinary differential equation (ODE)**, for $t \in [0, \infty)$

$$\boxed{\frac{ds}{dt}(t) = f_c(s(t), a(t))} \quad \text{with "right hand side function" } f_c : \mathbb{S} \times \mathbb{A} \rightarrow \mathbb{R}^{n_s}$$

(stochastic continuous time systems need intricate notation and are therefore omitted here)

Notation for Ordinary Differential Equation (ODE) Models



- ▶ denote $\frac{ds}{dt}(t)$ by $\dot{s}(t)$
- ▶ drop time argument, abbreviate $\dot{s}(t) = f_c(s(t), a(t))$ by

$$\dot{s} = f_c(s, a)$$

- ▶ In this course, we use the RL notation: s for state and a for control action
- ▶ But in control engineering, one uses: x for state and u for control action, i.e.,

$$\dot{x} = f_c(x, u)$$

(this notation might accidentally "slip through" on some slides)



ODE Example: Harmonic Oscillator

Mass m with spring constant k and friction coefficient β :

$$\begin{aligned}\dot{x}_1(t) &= x_2(t) \\ \dot{x}_2(t) &= -\frac{k}{m}(x_2(t) - u(t)) - \frac{\beta}{m}x_1(t)\end{aligned}$$

- state $x(t) \in \mathbb{R}^2$
- position of mass $x_1(t)$ ← measured
- velocity of mass $x_2(t)$
- control action: spring position $u(t) \in \mathbb{R}$ ← manipulated

Can summarize as $\dot{x} = f_c(x, u)$ with

$$f_c(x, u) = \begin{bmatrix} x_2 \\ -\frac{k}{m}(x_2 - u) - \frac{\beta}{m}x_1 \end{bmatrix}$$



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$$f_c(s, a) = \begin{bmatrix} s_2 \\ -\frac{k}{m}(s_2 - a) - \frac{\beta}{m}s_1 \end{bmatrix}$$

Some ODE Examples - what are their state vectors?



- ▶ Pendulum
- ▶ Hot plate with pot
- ▶ Continuously Stirred Tank Reactors (CSTR)
- ▶ Robot arms
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From Continuous to Discrete Time via Direct Multiple Shooting

[Bock & Plitt 1984]

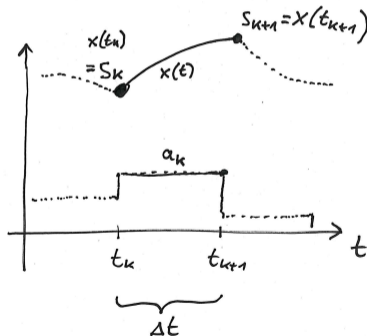


Transform continuous $\dot{s}(t) = f_c(s(t), a(t))$ into discrete time $s_{k+1} = f(s_k, a_k)$ as follows:

1. define $s_k := s(t_k)$ on **equidistant time grid** $t_k = k \Delta t$ with sampling time Δt
2. use **zero order hold** control $a(t) = a_k$ on $t \in [t_k, t_{k+1}]$
3. use **numerical simulation** to compute ODE solution $x(t) \equiv x(t; s_k, a_k)$ satisfying

$$\begin{aligned}x(t_k) &= s_k \\ \dot{x}(t) &= f_c(x(t), a_k) \quad \text{for } t \in [t_k, t_{k+1}]\end{aligned}$$

4. define $f(s_k, a_k) := x(t_{k+1}; s_k, a_k)$



From Continuous to Discrete Time via Direct Multiple Shooting

[Bock & Plitt 1984]



Transform continuous $\dot{s}(t) = f_c(s(t), a(t))$ into discrete time $s_{k+1} = f(s_k, a_k)$ as follows:

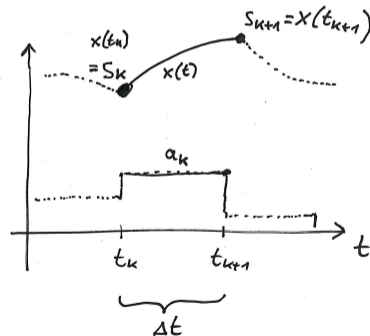
Exact ODE solution

$$x(0) = s,$$

$$\dot{x}(t) = f_c(x(t), a),$$

for $t \in [0, \Delta t]$

$$f(s, a) := x(\Delta t)$$



How to simulate ODE numerically?

Numerical Simulation/Integration, Three Examples

- ▶ simplest (but not recommended) implementation is a single step of an Euler integrator:

$$f(s, a) := s + \Delta t f_c(s, a)$$

- ▶ more accurate are N steps of an Euler integrator:

$$x_0 := s$$

for $i = 0$ to $N - 1$ do

$$x_{i+1} := x_i + (\Delta t/N) f_c(x_i, a)$$

$$f(s, a) := x_N$$

- ▶ more efficient are higher order **Runge Kutta (RK)** methods, e.g. a single RK4 step:

$$v_1 := f_c(s, a)$$

$$v_2 := f_c(s + (\Delta t/2) v_1, a)$$

$$v_3 := f_c(s + (\Delta t/2) v_2, a)$$

$$v_4 := f_c(s + \Delta t v_3, a)$$

$$f(s, a) := s + (\Delta t/6) (v_1 + 2v_2 + 2v_3 + v_4)$$



Euler vs 4th Order Runge Kutta Method (RK4) for Test Problem

Aim: solve $\dot{s} = s + a$ for $\Delta t = 1, s = 1, a = 0$. Exact solution is $f(s, a) = e = 2.718$.

► Four Euler steps give

$$x_0 := 1$$

$$x_1 := x_0 + 1/4x_0 \quad [= (1 + 1/4)x_0]$$

$$x_2 := (1 + 1/4)x_1$$

$$x_3 := (1 + 1/4)x_2$$

$$x_4 := (1 + 1/4)x_3$$

$$f_{\text{Euler}}(s, a) := x_4 \quad [= (1+1/4)^4 = 2.441], \text{ error} > 10\%$$

► One RK4 step gives

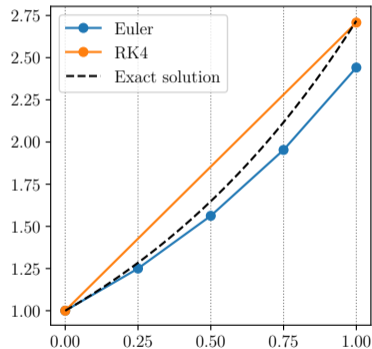
$$v_1 := 1$$

$$v_2 := 1 + 1/2v_1 \quad [= 6/4]$$

$$v_3 := 1 + (1/2)v_2 \quad [= 7/4]$$

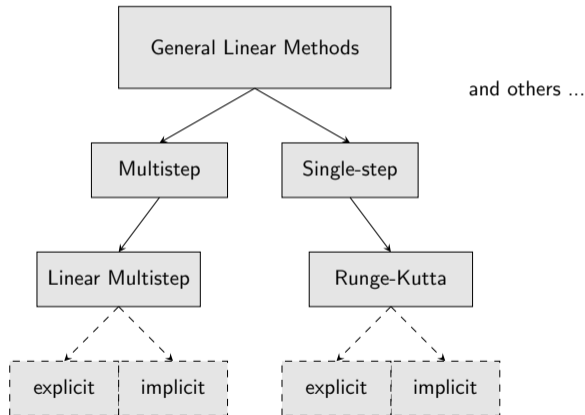
$$v_4 := 1 + v_3 \quad [= 11/4]$$

$$f_{\text{RK4}}(s, a) := 1 + (1/6)(v_1 + 2v_2 + 2v_3 + v_4) \quad [= 2.708]$$



RK4 is 27x more accurate than Euler for same number $M = 4$ of function evaluations

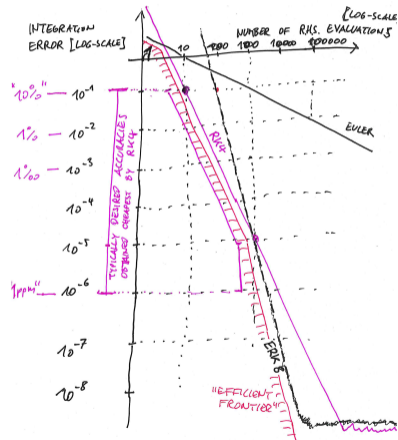
Classes of Numerical Simulation Methods



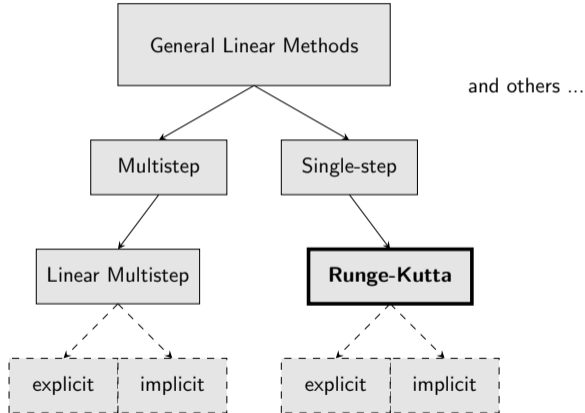
Fourth order RK method most efficient for typically desired accuracies



- ▶ each integration method is characterized by
 - ▶ integration order P and
 - ▶ number of internal stages S
- ▶ can increase accuracy by more integration steps N
- ▶ total number of function evaluations is $M = N \cdot S$
- ▶ integration error proportional to M^{-P}
- ▶ for small M , low order methods are most accurate, e.g., Euler with $P = 1$
- ▶ for large M , high order methods are more accurate
- ▶ humans typically want errors smaller than 10%, but rarely smaller than 10^{-6}
- ▶ accidentally, this favours the RK4 method ($P = 4$)



Classes of Numerical Simulation Methods



Discretization equations for general Runge Kutta (RK) methods

Exact ODE solution

$$x(0) = \mathbf{s},$$

$$\dot{x}(t) = v(t)$$

$$v(t) = f_c(x(t), \mathbf{a}),$$

$$\text{for } t \in [0, \Delta t]$$

$$f(\mathbf{s}, \mathbf{a}) := x(\Delta t)$$

N steps of general RK method with S stages

$$x_0 = \mathbf{s}, \quad x_{k+1} = x_k + h \sum_{j=1}^S b_j v_{k,j}$$

$$x_{k,i} = x_k + h \sum_{j=1}^S a_{ij} v_{k,j}$$

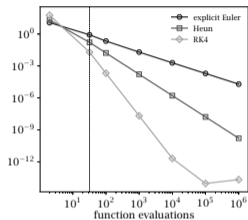
$$v_{k,i} = f_c(x_{k,i}, \mathbf{a}),$$

$$\text{for } i = 1, \dots, S, \quad k = 0, \dots, N-1$$

$$f(\mathbf{s}, \mathbf{a}) := x_N$$

- ▶ a_{ij} and b_j are **Butcher tableau entries** of (potentially implicit) Runge Kutta method
- ▶ step length $h := \Delta t/N$; intermediate states $x_k, x_{k,i}, v_{k,i} \in \mathbb{R}^{n_s}$ with integration step index $k \in \{0, 1, \dots, N\}$ and RK stage index $i, j \in \{1, \dots, S\}$
- ▶ N nonlinear equation systems with each $2Sn_s$ equations in $2Sn_s$ unknowns ($x_{k,i}, v_{k,i}$)
- ▶ solved by Newton's method (or imposed as equality constraints in optimization)

Butcher Tableau, Six Examples



Euler

$$\begin{array}{c|c} 0 & \\ \hline & 1 \end{array}$$

Heun

$$\begin{array}{c|cc} 0 & & \\ \hline 1 & 1 & \\ \hline & 1/2 & 1/2 \end{array}$$

RK4

$$\begin{array}{c|cccc} 0 & & & & \\ \hline 1/2 & 1/2 & & & \\ 1/2 & 0 & 1/2 & & \\ 1 & 0 & 0 & 1 & \\ \hline & 1/6 & 2/6 & 2/6 & 1/6 \end{array}$$

Implicit Euler

$$\begin{array}{c|c} 1 & 1 \\ \hline & 1 \end{array}$$

Midpoint rule (GL2)

$$\begin{array}{c|cc} 1/2 & 1/2 \\ \hline & 1 \end{array}$$

Gauss-Legendre of order 4 (GL4)

$$\begin{array}{c|cc} 1/2 - \sqrt{3}/6 & 1/4 & 1/4 - \sqrt{3}/6 \\ 1/2 + \sqrt{3}/6 & 1/4 + \sqrt{3}/6 & 1/4 \\ \hline & 1/2 & 1/2 \end{array}$$

$$\begin{array}{c|cccc} c_1 & a_{11} & \cdots & a_{1s} \\ c_2 & a_{21} & \cdots & a_{2s} \\ \vdots & \vdots & & \vdots \\ c_s & a_{s1} & \cdots & a_{ss} \\ \hline & b_1 & \cdots & b_s \end{array}$$

Intermediate Milestone: Deterministic State Space Models

From now on, throughout the course, we exclusively focus on discrete time models

$$s_{k+1} = f(s_k, a_k)$$

with integer time index $k = 0, 1, 2, \dots$. We often simplify notation to

$$\boxed{s^+ = f(s, a)}$$

Aim of optimal feedback control (including both MPC and RL) is to design a map, or **policy**, $\pi: \mathbb{S} \rightarrow \mathbb{A}$, $s \mapsto a := \pi(s)$ such that **closed-loop system** $\boxed{s^+ = f(s, \pi(s))}$ has desirable properties, such as respecting constraints and minimizing a cost.

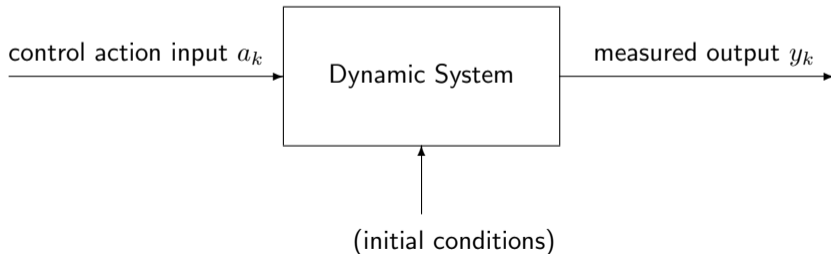
In practice, however, we might not be able to directly measure the state s ...



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The (realistic) Input Output Perspective

- ▶ In practice, we cannot measure the state. And the state representation is not even unique.
- ▶ A system model should allow us to predict, for any horizon length N and sequence of control actions (a_1, \dots, a_N) , the sequence of measured **outputs** (y_0, \dots, y_N) .
- ▶ Typically, we need to also specify some **initial conditions** (e.g. the initial state s_0)



Two Ways to Represent Deterministic Systems with Outputs

- ▶ State Space Models with outputs:

$$\begin{aligned} s_{k+1} &= f(s_k, a_k) \\ y_k &= g(s_k, a_k) \end{aligned} \quad \text{for } k = 0, 1, 2, \dots$$

Initial conditions = initial state s_0 .

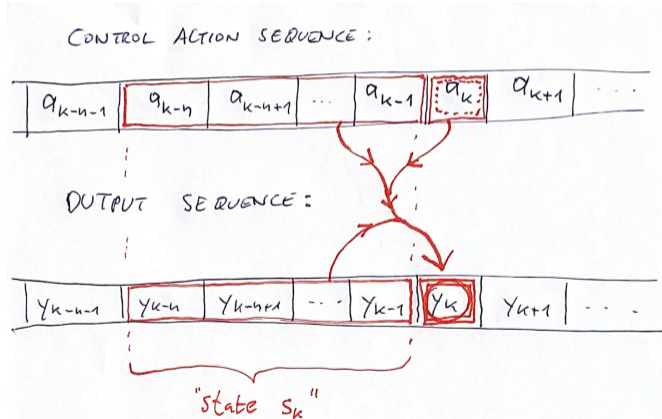
- ▶ Input Output Models (of order n):

$$y_k = h(y_{k-1}, \dots, y_{k-n}, a_k, \dots, a_{k-n}) \quad \text{for } k = n, n+1, n+2, \dots$$

Initial conditions: y_0, \dots, y_{n-1} and a_0, \dots, a_{n-1} .

Recurrence Equation in Input Output Models of order n

Visualization of recurrence $y_k = h(y_{k-1}, \dots, y_{k-n}, a_k, a_{k-1}, \dots, a_{k-n})$:



State Space Form of Input Output Models

- ▶ can **always transform input-output to state-space models**:
- ▶ state: $s_k = (y_{k-1}, a_{k-1}, \dots, y_{k-n}, a_{k-n})$ (defined for $k \geq n$)
- ▶ state transition $s \mapsto s^+ = f(s, a)$ described by

$$s_k = \begin{bmatrix} y_{k-1} \\ a_{k-1} \\ \vdots \\ y_{k-n+1} \\ a_{k-n+1} \\ y_{k-n} \\ a_{k-n} \end{bmatrix} \mapsto s_{k+1} = \begin{bmatrix} y_k \\ a_k \\ y_{k-1} \\ a_{k-1} \\ \vdots \\ y_{k-n+1} \\ a_{k-n+1} \end{bmatrix} = f(s_k, a_k) := \begin{bmatrix} h(y_{k-1}, \dots, y_{k-n}, a_k, \dots, a_{k-n}) \\ a_k \\ y_{k-1} \\ a_{k-1} \\ \vdots \\ y_{k-n+1} \\ a_{k-n+1} \end{bmatrix}$$

- ▶ output equation: $y_k = g(s_k, a_k) := h(y_{k-1}, \dots, y_{k-n}, a_k, \dots, a_{k-n})$.
- ▶ conversely, we can **sometimes transform state-space to input-output models**, e.g. in case of observable and controllable linear time invariant (LTI) models

Linear Time Invariant (LTI) Input Output Models

- ▶ Difference equation for **Auto Regressive models with eXogenous inputs (ARX)**:

$$y_k = c_1 y_{k-1} + \dots + c_n y_{k-n} + b_0 a_k + \dots + b_n a_{k-n}$$

for $k = n, n + 1, \dots$, with initial conditions: y_0, \dots, y_{n-1} and a_0, \dots, a_{n-1} .

- ▶ also called **Infinite Impulse Response (IIR)** model (if some c_i coefficients are nonzero)
- ▶ If all $c_i = 0$ we speak of **Finite Impulse Response (FIR)** models:

$$y_k = b_0 a_k + \dots + b_n a_{k-n}$$

- ▶ There exist also auto regressive (AR) models without inputs:

$$y_k = c_1 y_{k-1} + \dots + c_n y_{k-n}$$

Example: Fibonacci numbers 1,1,2,3,5,8,13,21, ... (with $c_1 = c_2 = 1$ and $y_0 = y_1 = 1$)

Some ODE Examples - what can be measured ?



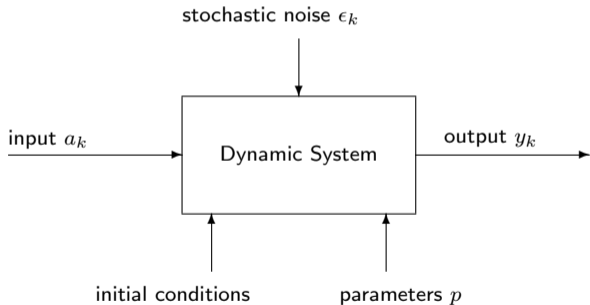
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- ▶ in reality, we always have some random noise ϵ_k , e.g., disturbances or measurement errors
- ▶ also, we usually have unknown, but constant system parameters p



(parameters can be seen as states that obey the dynamics $p_{k+1} = p_k$ and will often be omitted)



General Form (with random ϵ_k):

Stochastic State Space Model

$$\begin{aligned}s_{k+1} &= f(s_k, a_k, \epsilon_k) \\ y_k &= g(s_k, a_k, \epsilon_k)\end{aligned}$$

Stochastic Input Output Model

$$y_k = h(y_{k-1}, \dots, y_{k-n}, a_k, \dots, a_{k-n}, \epsilon_k, \dots, \epsilon_{k-n})$$

Special Cases:

- ▶ State Noise and Output Errors:

$$\begin{aligned}s_{k+1} &= f(s_k, a_k) + \epsilon_k^{\text{SN}} \\ y_k &= g(s_k, a_k) + \epsilon_k^{\text{OE}}\end{aligned}$$

- ▶ Equation Errors:

$$y_k = h(y_{k-1}, \dots, y_{k-n}, a_k, \dots, a_{k-n}) + \epsilon_k^{\text{EE}}$$

(note: different than output error)



Prior to implementing an MPC controller, one needs to address two tasks:

► **System Identification (offline):**

use a long sequence of recorded input and output data, (a_0, \dots, a_N) and (y_0, \dots, y_N) , to identify parameters p using e.g. least squares optimization or subspace identification

► **State Estimation (online):**

estimate the state s_k by using the previous control actions $(\dots, a_{k-2}, a_{k-1})$ and the past measurements $(\dots, y_{k-2}, y_{k-1})$ using e.g. Extended Kalman Filter (EKF) or moving horizon estimation (MHE) (MHE uses a fixed window of past data for fitting)

Learning-based MPC typically refers to an online model adaptation, i.e., to estimating parameters online (for which MHE is particularly suitable) ("learning a model" = "system identification")

Note: need state estimation only for *partially observable markov decision processes (POMDP)*



State Space View:

Partially Observable MDP

$$s_{k+1} = f(s_k, a_k, \epsilon_k)$$
$$y_k = g(s_k, a_k, \epsilon_k)$$

with independent identically distributed ϵ_k

Fully Observable MDP

$$s_{k+1} = f(s_k, a_k, \epsilon_k)$$
$$y_k = s_k$$

with $y_k \in \mathbb{R}^{n_s}$

Probabilistic View:

Partially Observable MDP

$$P_{\text{state}}(s_{k+1} | s_k, a_k)$$
$$P_{\text{meas}}(y_k | s_k, a_k)$$

with probability density functions $P(\cdot)$

Fully Observable MDP

$$P_{\text{state}}(s_{k+1} | s_k, a_k)$$
$$P_{\text{meas}}(y_k | s_k, a_k) = \delta(y_k - s_k)$$

with Dirac's Delta function $\delta(\cdot)$ in \mathbb{R}^{n_s}



Input output (I/O) models avoid need for state estimation

- ▶ We can avoid estimation task by assuming input-output (I/O) models of fixed order n
- ▶ This assumption leads to a **fully observable** markov decision process (MDP)
- ▶ State s_k at time k is then given by $s_k = (y_{k-1}, a_{k-1}, \dots, y_{k-n}, a_{k-n})$
- ▶ Reinforcement Learning (RL) algorithms often use I/O-models ("end-to-end learning")
- ▶ I/O-models also used in some *linear MPC* implementations based on LTI models, e.g.

$$y_k = \sum_{i=0}^n b_i a_{k-i} + (y_{k-1} - \sum_{i=0}^n b_i a_{k-i-1}) + \epsilon_k$$

- ▶ I/O-models also used for nonlinear black-box MPC or model-based RL which use neural networks for the mapping $y_k = h(y_{k-1}, \dots, y_{k-n}, a_k, \dots, a_{k-n})$



- ▶ We distinguish different model types
 - ▶ continuous vs discrete state and control
 - ▶ continuous vs discrete time
 - ▶ linear vs nonlinear
 - ▶ state space vs input output
 - ▶ deterministic vs stochastic
 - ▶ fully or partially observable
(not to be confused with "observability" in systems theory)
- ▶ We transform differential equations to discrete time via *numerical simulation*
- ▶ We denote *deterministic discrete time models* and *Markov Decision Processes (MDP)* by

$$\boxed{s^+ = f(s, a)} \quad \text{and} \quad \boxed{P(s^+ | s, a)}$$

with state $s \in \mathbb{R}^{n_s}$ and control action $a \in \mathbb{R}^{n_a}$