

## Exercise Sheet 1 with solutions

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Within the control part of the "Energy Systems: Hardware and Control" course there will be 45 min. exercise sessions after each lecture. The exercises are guided by tutors and will contain some **MATLAB**-based tasks. Therefore, a **MATLAB** installation including the Control System Toolbox is needed.

### Getting started

1. Except **MATLAB** is not yet installed on your computer, the first thing you need to do is install it. Detailed installation and licensing instructions can be found at <https://www.rz.uni-freiburg.de/services-en/beschaffung-em/software-en/matlab-license>  
Remember that the Control System Toolbox is required.
2. If you are new to **MATLAB**, the first thing you will appreciate is the extensive help system. You can simply type `doc` into the console and the documentation opens. If you type `doc plot`, you will find a detailed description of function `plot`.
3. Here are some useful commands for the exercises:  

```
hold on/off  
figure  
close all  
clear  
clc
```

### Problem 1: Dynamical System, ODE, Simulation and Solution

A simple pendulum is sketched in figure 1. The point-mass  $m$  is fixed to a solid, massless rod of length  $l$ , which is connected to a frictionless hinge on the other side. All movements take place in the vertically oriented x-y-plane and the gravitation  $g$  acts in y-direction.

- (a) Derive the equation of motion for the pendulum and note it in the shape  $\ddot{\alpha} = f(\alpha)$ . How do the mass or the length determine the motion of the pendulum?

$$E = \frac{1}{2}ml^2\dot{\alpha}^2 + mgl(1 - \cos(\alpha))$$
$$\frac{dE}{dt} = 0 \quad \text{conservation of energy}$$
$$\ddot{\alpha} = -\frac{g}{l} \sin(\alpha)$$

- (b) What are the states  $x$ , which are needed to completely describe the system?

$$x = \begin{bmatrix} \alpha \\ \dot{\alpha} \end{bmatrix} = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

- (c) Convert the ODE to the system of equations  $\dot{x} = f(x)$ .

$$\dot{x} = \begin{bmatrix} \dot{\alpha} \\ \ddot{\alpha} \end{bmatrix} = \begin{bmatrix} \dot{\alpha} \\ -\frac{g}{l} \sin(\alpha) \end{bmatrix} = \begin{bmatrix} x_2 \\ -\frac{g}{l} \sin(x_1) \end{bmatrix}$$

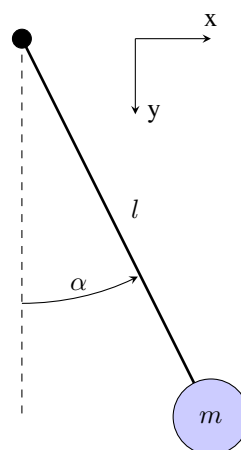


Figure 1: Sketch of a simple pendulum

- (d) Simulate the motion of the pendulum for 10 seconds using different initial values  $x_0$ . Therefore write a function `dx = nonlin_pendel(t, x)` which implements the system of equations. For the simulation use `lsode` and the following constants:  $l = 1 \text{ m}$ ,  $g = 9.81 \frac{\text{m}}{\text{s}^2}$ . Solution see `problem1.m`.
- (e) What characterizes steady states? Calculate the steady states for the pendulum.

$$\begin{aligned} \dot{x} &\stackrel{!}{=} 0 \\ 0 &= \begin{bmatrix} x_2 \\ -\frac{g}{l} \sin(x_1) \end{bmatrix} \\ \Rightarrow x_{2\text{ss}} &= 0, \quad x_{1\text{ss}} = n \cdot \pi \quad \text{with } n \in \mathbb{Z} \\ x_{\text{ss}} &= \begin{bmatrix} n \cdot \pi \\ 0 \end{bmatrix} \end{aligned}$$

- (f) Linearize the system at the steady state  $x_{\text{ss}} = [0, 0]^\top$  and write the system of equations in the shape  $\dot{x} = Ax + Bu$  and  $y = \alpha + Cx$ . Compute the state space matrices  $A, B$  and  $C$ .

$$\begin{aligned} \dot{x} &= \begin{bmatrix} \frac{\partial f_1}{\partial x_1}(x_{\text{ss}}) & \frac{\partial f_1}{\partial x_2}(x_{\text{ss}}) \\ \frac{\partial f_2}{\partial x_1}(x_{\text{ss}}) & \frac{\partial f_2}{\partial x_2}(x_{\text{ss}}) \end{bmatrix} x \\ &= \begin{bmatrix} 0 & 1 \\ -\frac{g}{l} \cos(x_{1\text{ss}}) & 0 \end{bmatrix} x = \begin{bmatrix} 0 & 1 \\ -\frac{g}{l} \cos(0) & 0 \end{bmatrix} x = \begin{bmatrix} 0 & 1 \\ -\frac{g}{l} & 0 \end{bmatrix} x \end{aligned}$$

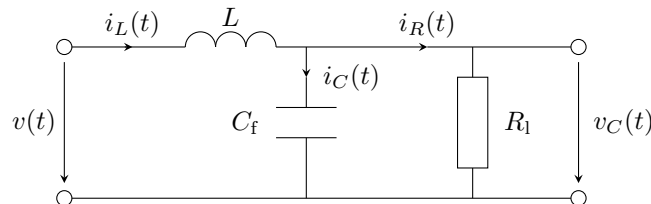
- (g) Compare the linear and the nonlinear system via simulations using increasing initial values for  $\alpha(0)$  ranging from  $\pi/8$  to  $\pi$ .

*With increasing  $\alpha(0)$  the linear approximation gets worse. But for control tasks we want to keep our system close to a desired state, so the linearization is a good simplification and thus an important and very useful tool. For plot run `problem1.m`*

## Problem 2: Buck-converter, Modelling and Stabilization

1. The electrical circuit sketched below shows a simplified buck-converter with a constant load at the output. The system can be described in state-space representation as

$$\begin{aligned} \dot{x} &= \mathbf{A}x + \mathbf{B}u, \quad y = \mathbf{C}x, \quad \mathbf{D} = [0], \\ u &:= v, \quad y := v_C. \end{aligned}$$



- (a) Derive the the I/O-ODE (Input/Output-Ordinary Differential Equation) for the given circuit using equations

$$i_C = C_f \frac{dv_C}{dt}, \quad v_L = L \frac{di_L}{dt} \quad \text{and} \quad i_R = \frac{v_C}{R_l}$$

(Hint: Use Kirchhoff's voltage law for inductors and current law for capacitors)

$$\begin{aligned} v &= v_C + v_L = v_C + L \cdot (\dot{i}_C + \dot{i}_R) \\ \ddot{v}_C + \frac{1}{R_l C_f} \dot{v}_C + \frac{1}{L C_f} v_C &= \frac{1}{L C_f} v \end{aligned}$$

- (b) Convert the I/O-ODE to state space representations, i.e. set up the  $\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D}$ -matrices for

$$\dot{x}(t) = \mathbf{A}x(t) + \mathbf{B}u(t) \quad (1)$$

$$y(t) = \mathbf{C}x(t) + \mathbf{D}u(t) \quad (2)$$

using

- (i) the control canonical form.

$$\dot{\mathbf{x}} = \begin{bmatrix} 0 & 1 \\ -\frac{1}{LC_f} & -\frac{1}{R_1 C_f} \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \mathbf{u}$$

$$y = \begin{bmatrix} \frac{1}{LC_f} & 0 \end{bmatrix} \mathbf{x}$$

(ii) the observer canonical form.

$$\dot{\mathbf{x}} = \begin{bmatrix} 0 & -\frac{1}{LC_f} \\ 1 & -\frac{1}{R_1 C_f} \end{bmatrix} \mathbf{x} + \begin{bmatrix} \frac{1}{LC_f} \\ 0 \end{bmatrix} \mathbf{u}$$

$$y = \begin{bmatrix} 0 & 1 \end{bmatrix} \mathbf{x}$$

(c) Now derive matrices  $\mathbf{A}$ ,  $\mathbf{B}$ , and  $\mathbf{C}$  for the state vector given as  $\mathbf{x} := [i_L \quad v_C]^\top$ .

Inductor current:

$$\frac{di_L}{dt} = \frac{v_L}{L}$$

Express  $v_C$  by utilizing Kirchoff's voltage law:

$$v_L = v - v_C$$

$$\Rightarrow \frac{di_L}{dt} = \frac{v}{L} - \frac{v_C}{L}$$

Capacitor voltage:

$$\frac{dv_C}{dt} = \frac{i_C}{C_f}$$

Express  $i_C$  by utilizing Kirchoff's current law:

$$i_C = i_L - i_R = i_L - \frac{v_C}{R_1}$$

$$\Rightarrow \frac{dv_C}{dt} = \frac{i_L}{C_f} - \frac{v_C}{C_f R_1}$$

State-space representation:

$$\dot{\mathbf{x}} = \begin{bmatrix} \frac{di_L}{dt} \\ \frac{dv_C}{dt} \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & -\frac{1}{L} \\ \frac{1}{C_f} & -\frac{1}{R_1 C_f} \end{bmatrix}}_{\mathbf{A}} \mathbf{x} + \underbrace{\begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix}}_{\mathbf{B}} \underbrace{v}_{\mathbf{u}}$$

$$\mathbf{C} = \begin{bmatrix} 0 & 1 \end{bmatrix}$$

(d) Derive the characteristic polynomial. Evaluate the eigenvalues of the system for  $L = 4.7$  mH,  $C_f = 100$   $\mu$ F and

- i.  $R_1 = \infty \Omega$
- ii.  $R_1 = 100 \Omega$

Is the system BIBO-stable in both cases?

Characteristic polynomial:

$$p(\lambda) = \det(\lambda \mathbf{I} - \mathbf{A})$$

$$= \det \begin{bmatrix} \lambda & \frac{1}{L} \\ -\frac{1}{C_f} & \lambda + \frac{1}{R_1 C_f} \end{bmatrix}$$

$$= \lambda^2 + \frac{\lambda}{R_1 C_f} + \frac{1}{LC_f}$$

Roots of the characteristic polynomial:

- $R_1 = \infty \Omega$ :

$$\lambda^2 + \frac{1}{LC_f} \stackrel{!}{=} 0$$

$$\Rightarrow \lambda = \pm \frac{i}{\sqrt{LC_f}} \approx \pm i 1.46 \cdot 10^3$$

$\text{Re}(\lambda_i) = 0$  for all eigenvalues.

$\Rightarrow$  system is undamped and therefore not BIBO-stable.

- $R_1 = 100 \Omega$

$$\lambda^2 + \frac{\lambda}{R_1 C_f} + \frac{1}{LC_f} \stackrel{!}{=} 0 \Rightarrow \lambda = -\frac{1}{2R_1 C_f} \pm i \sqrt{-\frac{1}{4R_1^2 C_f^2} + \frac{1}{LC_f}} \approx -50 \pm i 1.46 \cdot 10^3$$

$\text{Re}(\lambda_i) < 0$  for all eigenvalues.

$\Rightarrow$  system is damped and therefore BIBO-stable.

(e) Write down the time constant  $\tau$  in seconds and the resulting oscillating frequency in Hz for both values of  $R_1$ .

(Hint: In this example, the time constant  $\tau$  is a measure for the amplitude decay (damping) and is defined as  $\tau = -\frac{1}{\text{Re}(\lambda)}$ . The oscillating frequency is defined as  $f_0 = \frac{\omega_0}{2\pi} = \frac{|\text{Im}(\lambda)|}{2\pi}$  )

- $R_1 = \infty \Omega$ :

Time constant  $\tau$ :

$$\tau = -\frac{1}{\text{Re}(\lambda)} = -\infty \text{ ms}$$

Oscillating Frequency:

$$\omega_0 = |\text{Im}(\lambda)| = \frac{1}{\sqrt{LC_f}}$$

$$f_0 = \frac{\omega_0}{2\pi} \approx 232.15 \text{ Hz}$$

- $R_1 = 100 \Omega$

Time constant  $\tau$ :

$$\tau = -\frac{1}{\text{Re}(\lambda)} = 2C_f R_1 = 20 \text{ ms}$$

Oscillating Frequency:

$$\omega_0 = |\text{Im}(\lambda)| = \sqrt{-\frac{1}{4R_1^2 C_f^2} + \frac{1}{LC_f}}$$

$$f_0 = \frac{\omega_0}{2\pi} \approx 232.01 \text{ Hz}$$

(f) Create a new MATLAB script and define variables  $L = 4.7 \text{ mH}$ ,  $C_f = 100 \mu\text{F}$  and  $R_1 = \infty$ . Also define matrices  $\mathbf{A}$ ,  $\mathbf{B}$ ,  $\mathbf{C}$  and  $\mathbf{D} = 0$  according to task (1c).

(g) Use the `ss(A,B,C,D)` command to create a state-space model `sys_01` and evaluate the systems step response with the `step(sys,Tfinal)` function ( $T_{\text{final}} = 0.1 \text{ s}$ ) for

- $R_1 = \infty \Omega$
- $R_1 = 100 \Omega$

(h) Is the system controllable and/or stabilizable?

$$\mathbf{C} = [\mathbf{B} \quad \mathbf{AB}] = \begin{bmatrix} \frac{1}{L} & 0 \\ 0 & \frac{1}{C_f L} \end{bmatrix}$$

$$\det(\mathbf{C}) = \frac{1}{C_f L^2} \neq 0$$

The system is fully controllable and therefore also stabilizable.

2. Now we want to introduce a state feedback with gain  $K$  to stabilize the system in the case where no load is connected ( $R_l = \infty$ ).

(a) Where do the two poles have to be shifted to obtain the following characteristics for the closed-loop system?:  
 $\tau = 10$  ms,  $f_0 = 100$  Hz

$$\begin{aligned}\operatorname{Re}(\lambda_{1/2}) &= -\frac{1}{\tau} = -100 \\ \operatorname{Im}(\lambda_{1/2}) &= \pm 2f_0\pi \approx \pm 628.3185\end{aligned}$$

(b) Use the MATLAB function `place(A, B, p)` to calculate the corresponding feedback vector  $K$  and implement the system matrix  $\mathbf{A}_{cl}$  as well as the closed-loop model `sys_stable` for the stabilized system.

$$\begin{aligned}\mathbf{A}_{cl} &= \mathbf{A} - \mathbf{B}K \\ (\mathbf{B}_{cl} &= \mathbf{B}, \mathbf{C}_{cl} = \mathbf{C})\end{aligned}$$

(c) Simulate `sys_stable` with the `step()` command and verify frequency and damping is as desired.